

Static active-matrix OLED display without pixel refresh enabled by amorphous-silicon non-volatile memory

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Abstract — An active-matrix organic light-emitting-diode (AMOLED) display which does not require pixel refresh is demonstrated. This was achieved by replacing the thin-film transistor (TFT) that drives the OLED with a non-volatile memory TFT, in a 2-transistor pixel circuit. The threshold voltage of the non-volatile-memory TFT can be changed by applying programming voltage pulses to the gate electrode. This approach eliminates the need for storage capacitors, increases the pixel fill factor, and potentially reduces power consumption. Each pixel can be individually programmed or erased using a standard active-matrix addressing scheme. The programmed image is stored in the display even if power is turned off.

Keywords — Amorphous silicon, thin-film transistor, non-volatile memory, AMOLED display.

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1 Introduction

There is growing interest in the development of novel AMOLED pixel circuitry to reduce power consumption and improve fill factor. Traditionally, AMOLED displays require periodic pixel refreshes, which consume power to recharge the storage capacitor even if the image is not changing. The storage capacitor also reduces pixel fill factor. In this work, we take advantage of the recent developments in a-Si non-volatile memory,¹⁻³ and demonstrate an AMOLED display that stores and displays images without pixel refreshes.

2 Background

The conventional AMOLED pixel [Fig. 1(a)] has two TFTs, a storage capacitor, and an OLED. During display operation, the select line is pulled high to allow the voltage on the data line to propagate through the switching TFT to the storage node, charging the storage capacitor and the gate electrode of the driver TFT. This sets the current through the driver TFT and OLED, and consequently the brightness of the pixel. The storage capacitor is necessary to hold the voltage on the gate of the driver TFT between pixel refreshes, but it also reduces the active light-emitting area of the pixel and thus the fill factor of the pixel. Furthermore, because the charge on the storage capacitor leaks through the switching TFT over time, the pixel must be refreshed (or recharged) periodically. In low-frame-rate applications, these excess refresh cycles are required to simply maintain the image, not to introduce new image data into the pixels.

We propose a new pixel structure [Fig. 1(b)] to address these two issues by integrating an a-Si memory TFT as the

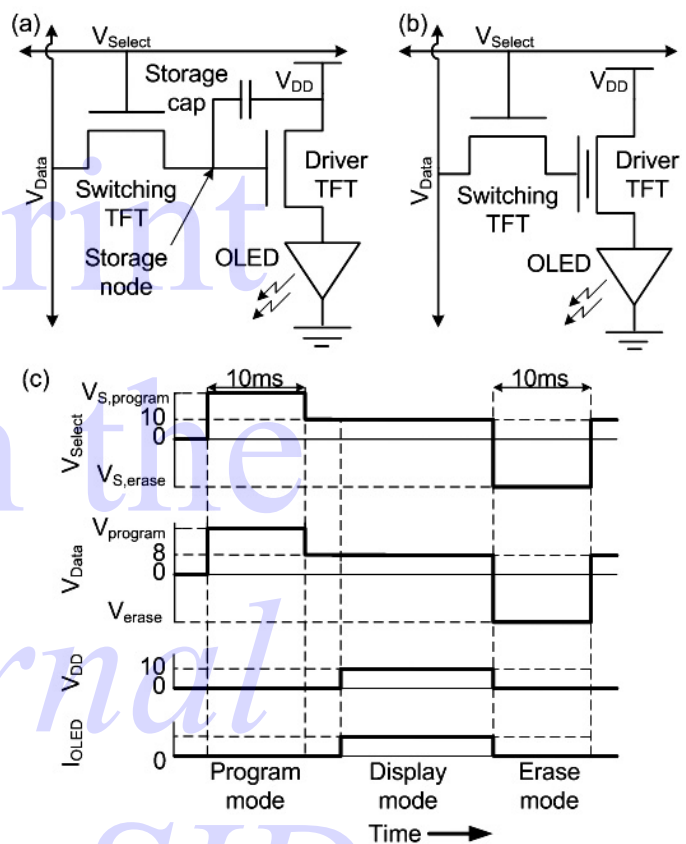


FIGURE 1 — (a) Conventional 2-TFT AMOLED pixel with the storage capacitor used to hold the voltage on the gate electrode of the driver TFT between pixel refreshes. (b) A 2-TFT AMOLED pixel with integrated memory device. (c) Timing diagram showing the modes of operation in the new pixel in (b).

driver TFT. These memory TFTs will be referred to as SiN_x trap TFTs (ST-TFTs) from here on. The ST-TFT has excellent memory characteristics, including a room-temperature

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retention time greater than 10 years.³ It uses a high-defect-density interface, embedded in the gate dielectric layer, to trap charges injected by short duration (typically 10 msec) voltage pulses applied to the gate. By varying the magnitude of the programming pulse, we can control the threshold voltage (V_T) of a ST-TFT. In the new pixel circuit with the ST-TFT driver, the OLED brightness is *not* controlled by varying the V_{DATA} applied to the gate of the driver ST-TFT through the switching TFT, as in a conventional AMOLED pixel. Instead, a constant V_{DATA} is applied to the gate of the driver ST-TFT. The current through the driver ST-TFT and the OLED brightness is controlled by changing the V_T of the ST-TFT through programming. If low brightness is desired, the ST-TFT is set to a high V_T by programming with a large positive gate pulse. If high brightness is desired, the ST-TFT is set to a low V_T by programming with a small positive gate pulse, or not programming at all and leaving it with the initial V_T .

A given pixel in a display is programmed by applying a high magnitude (*e.g.*, 35 V) and short duration (*e.g.*, 10 msec) voltage pulse to the corresponding data line [program mode in Fig. 1(c)]. The same voltage pulse (plus ~ 2 V to account for threshold voltage of the switching TFT) is applied to the corresponding select line to allow the V_{DATA} pulse to propagate to the gate of the desired driver ST-TFT. All other select lines are held at ground to prevent the V_{DATA} pulse from programming other undesired pixels. V_{DD} is held at ground during programming to ensure both the source and the drain of the ST-TFT are at 0 V. This process is repeated until all pixels in the display have been programmed to their respective brightness values.

After programming, the display is activated by setting V_{DD} to 10 V [display mode in Fig. 1(c)], V_{DATA} to 8 V on all the data lines, and V_{SELECT} to 10 V on all the select lines. The OLED current and therefore brightness of any given pixel is determined by the programmed V_T of the driver ST-TFT. High V_T translates into a small OLED current and a dim pixel, and a low V_T translates into large OLED current and a bright pixel. It is important to note that both V_{DATA} and V_{SELECT} are DC voltages in the display mode because a pixel refresh is not necessary to maintain a static image. The image information remains stored in the ST-TFT threshold voltage even if the power is turned off.

To change the programmed image, the pixels are first erased and then programmed again. Erase mode, shown in Fig. 1(c), is identical to the program operation. The only difference is that the applied voltage pulse has a large negative magnitude, instead of a positive one. This negative voltage forces the trapped electrons in the ST-TFT to tunnel back out, causing V_T to shift towards its original unprogrammed value. Note that to erase a single pixel in the active matrix (instead of an entire column), all other select lines would have to be held at -30 V to prevent the erase pulse from propagating to the undesired pixel drivers.

3 Fabrication

The a-Si ST-TFT and the AMOLED backplane were fabricated using a process that is identical to that of the conventional bottom-gate a-Si TFT, except for one key step – the formation of the SiN_x traps. After the control gate dielectric in the ST-TFT is deposited using plasma-enhanced chemical vapor deposition (PECVD), reactive ion etch (RIE) is used to create etch damage on the SiN_x surface. The etch damage results in midgap trap states, which are the workhorse of the ST-TFT memory. The traps are then encapsulated with the tunnel SiN_x , and the process follows that of a conventional bottom gate a-Si TFT from this point forward. More details of the a-Si ST-TFT and AMOLED backplane fabrication process are described in Refs. 3–5. After the completion of the backplane, which consists of the TFTs and the interconnect lines, the sample is encapsulated with a 300-nm-thick SiN_x layer deposited via PECVD. This layer serves as insulation between the backplane and the OLEDs. Via holes are etched into the passivation SiN_x using RIE to allow contact between the source of the driver TFT and indium–tin–oxide (ITO), which is the anode of the OLEDs. A 250-nm-thick ITO layer was deposited via RF sputtering with a gas mixture of Ar/O_2 (99%/1%) at room temperature. The ITO layer is patterned via wet etching with aqua regia ($\text{HNO}_3/\text{HCl}/\text{H}_2\text{O}$ 1:5:6). A 1.4- μm -thick photo-patternable planarization resist is spin-coated over the entire sample, and patterned to reveal the flat portions of the ITO anodes. The planarization resist is reflowed at 180°C to provide smooth edges that cover the rough features of the circuits and interconnect below. This improves the yield of the OLEDs by reducing the chance of short-circuit faults. The sample is exposed to UV-ozone for about 5 minutes, to increase the surface work function of the ITO anode and improve hole-injection efficiency.⁶ An organic multilayer of *N,N'*-diphenyl-*N,N'*-bis(3-methylphenyl)-1,1'-biphenyl-4,4'-diamine (TPD)/aluminum tris(8-hydroxyquinoline) (Alq3) was deposited via thermal evaporation to form green luminescent diodes.⁷ The thicknesses of both layers were ~ 30 nm. A bilayer of Mg/Ag (20 nm/100 nm) is evaporated via thermal evaporation to form the cathode of the OLEDs.

The cross section of the completed pixel is shown in Fig. 2. The pixel area is 0.2 mm². The switching TFT is 5 μm /5 μm and the driver TFT is 150 μm /15 μm . With a design rule of 10 μm , the fill factor is 81%, a 15% improvement over the 66% fill factor of a conventional pixel fabricated with the same design rule and TFT dimensions.

4 Results and discussion

The pixel current–voltage characteristics were first measured prior to OLED integration by applying a V_{DD} of 10 V, a V_{SELECT} of 10 V, grounding the source of the driver ST-TFT (OLED anode) and sweeping V_{DATA} from 0 to 8 V in 50-mV increments. V_{DATA} should be fully transferred to the gate voltage of the ST-TFT for slow scans, and with initial

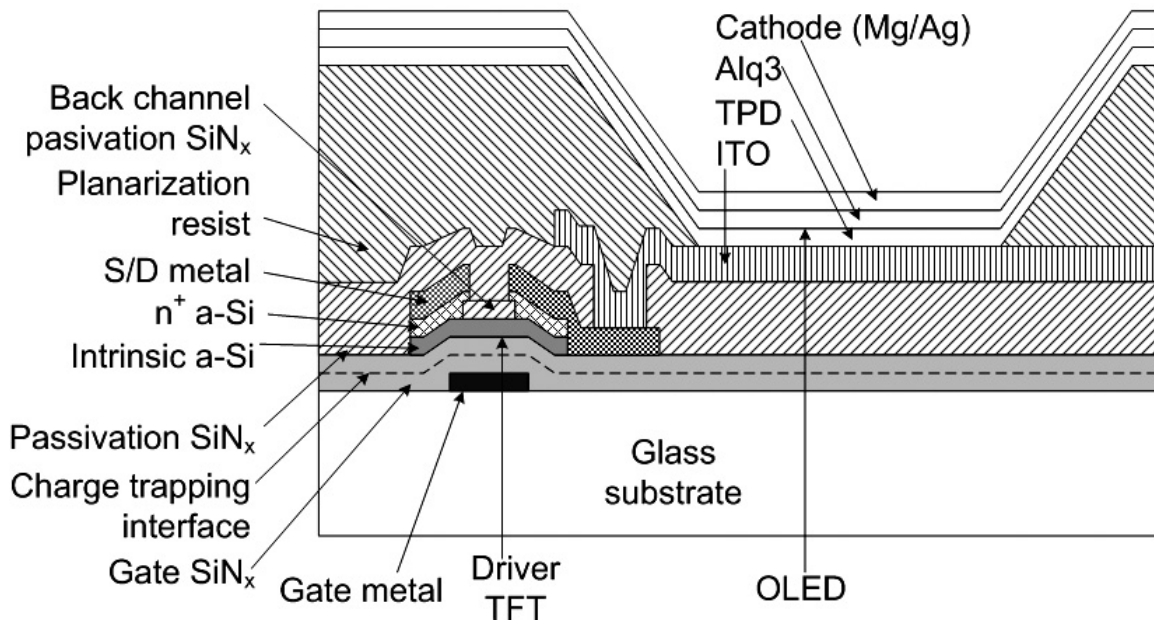


FIGURE 2 — Cross-sectional structure of AMOLED pixel with integrated a-Si non-volatile memory TFT.

TFT threshold voltage of ~ 1.5 V, the TFTs are in saturation mode. The curve marked “before program” in Fig. 3 is the pixel current *vs.* V_{DATA} characteristic of the as-fabricated pixel, and the one marked “after program at 35 V” is the pixel characteristic after a 10 msec 35-V programming pulse (with $V_{SELECT} = 37$ V) has been applied to the data lines with V_{DD} grounded. A clear V_T shift of ~ 2 V is observed. Consequently, under the same bias conditions, the programmed pixel would provide less drive current than the unprogrammed pixel, resulting in a dimmer OLED. The curve marked as “after erase” is the pixel characteristic after

the programmed pixel has been erased with a 10-msec -28.5 -V voltage pulse on V_{DATA} ($V_{SELECT} = -26.5$). It can be seen that the programming-induced V_T shift can be reversed *via* the erase operation and the pixel characteristics returned that of the unprogrammed state. More detailed characterization of the ST-TFT can be found elsewhere.³

Figure 4 shows the pixel brightness in an isolated test pixel as a function of programming voltage, after OLED integration with the cathode grounded. The pixel brightness is measured with fixed $V_{DATA} = 8$ V, $V_{DD} = 10$ V, and $V_{SELECT} = 10$ V. The measured pixel brightness decreases with increasing magnitude of the programming voltage pulse, with the pixel turned completely off for a programming voltage

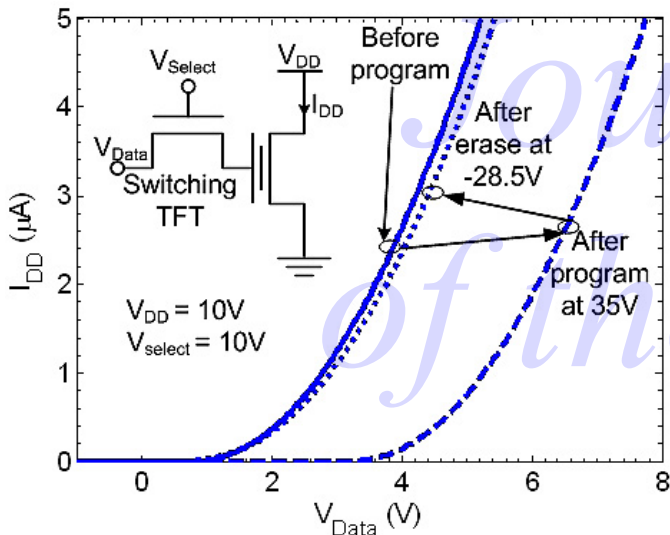


FIGURE 3 — Single-pixel current *vs.* V_{DATA} characteristics before OLED integration. The aspect ratios of the driver TFT and switching TFT are 10 and 1, respectively. The curves are shown for pixel before programming, after programming at 35 V ($V_{SELECT} = 37$ V) for 10 msec, and after erasing at -28.5 V ($V_{SELECT} = -26.5$ V) for 10 msec.

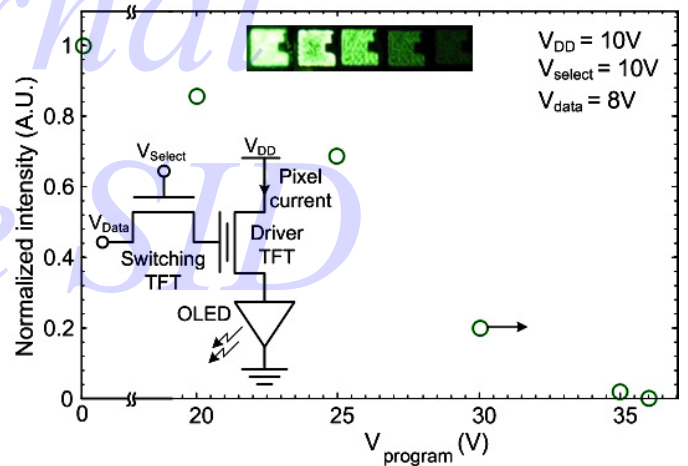


FIGURE 4 — Single pixel brightness as a function of programming voltage, after OLED integration. The inset shows photographs of five individual pixels, corresponding in order to the programming conditions of 0, 20, 25, 30, and 35 V. The aspect ratios of the driver TFT and switching TFT are 10 and 1, respectively.

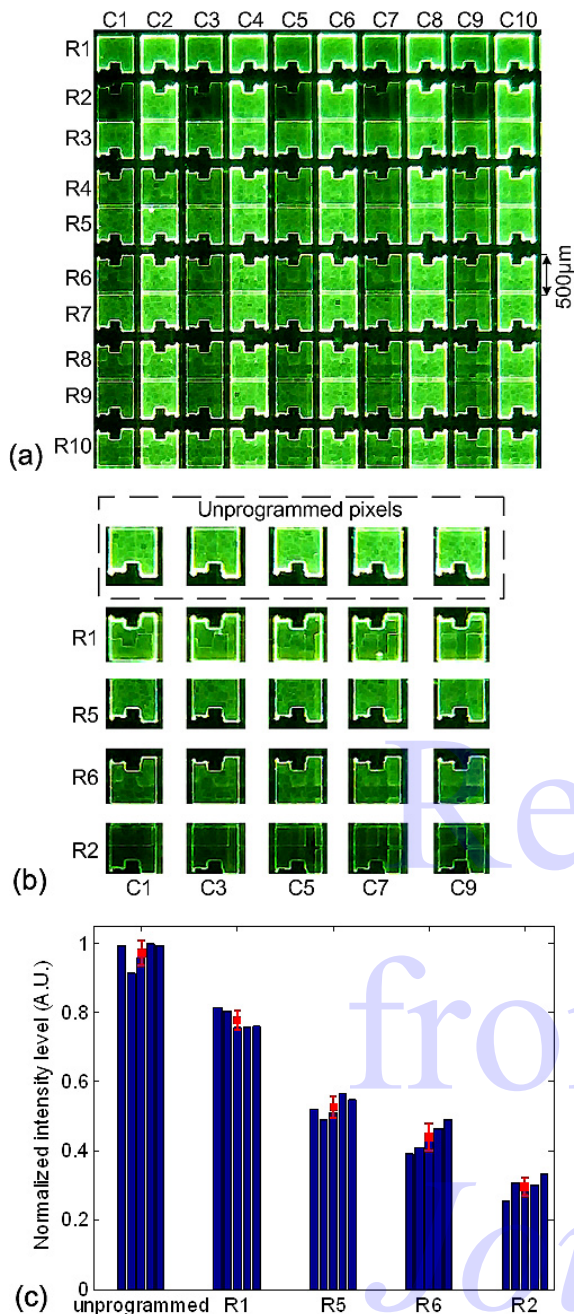


FIGURE 5 — (a) Photograph of an integrated 10×10 AMOLED display. The even columns (C2, C4, C6, C8, and C10) have not been programmed. The odd columns (C1, C3, C5, C7, and C9) are tied together and have been programmed to four different intensity levels, with programming voltages of 20 V (R1, R3 and R10), 24 V (R5 and R7), 28 V (R4, R6, R8 and R9), and 32 V (R2). (b) Representative pixels from each of the four different programmed intensity levels, along with the unprogrammed pixels (from the even columns), are reproduced to for better viewing comparison. (c) Measured brightness statistics of the pixels from the array.

of 37 V. The trend is in agreement with the expectation that as the threshold voltage shift of the driver ST-TFT increases with increasing programming voltages.

Figure 5(a) shows the photograph of an integrated 10×10 AMOLED display. All the even data lines (columns) are tied together, and all the odd data lines are tied together for simplified testing. The select lines (rows) can be accessed

individually. The odd columns were programmed by applying programming voltage pulses to the odd data lines, and no programming was done on the even columns. Programming was performed one row at a time, by activating the row select lines with the appropriate voltage pulse. Because the odd column data lines were tied together, all the odd-column pixels in a given row are programmed to the same intensity. The rows were programmed to four different intensity levels with programming voltages of 20, 25, 27, and 30 V. After programming, the display was driven with constant DC voltages of $V_{DD} = 10$ V, $V_{SELECT} = 10$ V, and $V_{DATA} = 8$ V. Representative pixels of each intensity level, along with unprogrammed pixels (from even rows), are reproduced in Fig. 5(b) for better brightness comparison. R1, R5, R6, and R2 are programmed with 20, 24, 28, and 32 V, respectively. As the programming voltage is increased, raising the ST-TFT threshold, the brightness of the pixel decreases as expected. Figure 5(c) shows the brightness-level statistics of pixels from an integrated AMOLED display. Pixels that have been programmed to the same brightness exhibit a standard deviation in brightness of $\sim 4\%$, demonstrating good uniformity and control in setting the pixel brightness.

5 Conclusion

We have successfully demonstrated a new approach for the control of pixel brightness in static and low-refresh-rate applications, by integrating a-Si non-volatile-memory TFTs into the pixel circuit. Because the display does not need storage capacitors, it achieves a better fill factor and does not need power-consuming refresh cycles. The brightness is controlled by adjusting the threshold voltage of the OLED driver TFT.

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