| 000 | ELE 301, Fall 2010 |
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| 001 002 | Laboratory No. 7 |
| 003 | Laboratory 100. 7 |
| 004 | Stability and Root Locus Plots |
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| 006 | |
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| 800 | 1 Background |
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| 010 011 | 1.1 Transfer Function of a Circuit |
| 012 | Model resistors, capacitors and inductors as linear time-invariant components. To obtain |
| 013 | the transfer function of a circuit made up of these components, we determine each individual |
| 014 | component's current to voltage transfer function (impedance). |
| 015 | The current-voltage relationship for a capacitance C is $I(t) = C(dV(t)/dt)$. So when $V(t) =$ |
| 016 | Ve^{st} , $I(t) = CsVe^{st} = Ie^{st}$. Hence $V = Z_C I$, where $Z_C = 1/(sC)$. |
| 017 | The current-voltage relationship for inductance is $V(t) = L(dI(t)/dt)$. As above consider |
| 018 | $V(t) = Ve^{st}$. Then $I(t) = LsVe^{st} = Ie^{st}$. So $V = Z_LI$, where $Z_L = sL$. You can apply the |
| 019 | same analysis to a resistance R to obtain $V = IR$. |
| 020 | So for inputs of the form e^{st} , circuit components have voltage-current relationships that |
| 021 | look like Ohm's law, except that the impedances are now complex: |
| 022 | |
| 023 024 | $Z_R = R$ $Z_C = 1/(sC)$ $Z_L = sL$ |
| 025 | All the rules of resistor combination in series and in parallel apply to impedances, so sinu- |
| 026 | soidal circuit analysis becomes very easy. |
| 027 | As an example, we analyze the RC circuits in Figure 1. We use basic circuit laws to calculate |
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| 030 | Vin R Vout Vin R Vout |
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| 037 | Figure 1: Two RC circuits. |
| 038 039 | |
| 040 | $v_{out} = H(s)v_{in}$. Each circuit is a voltage divider. For the circuit on the left: |
| 041 | 1/cC 1 |
| 042 | $v_{out} = v_{in} \frac{1/sC}{R+1/sC} = v_{in} \frac{1}{1+sRC}$ |
| 043 | n + 1/sU $1 + s$ nU |
| 044 | Hence |
| 045 | $H_1(s) = \frac{1}{1 + sCR}$ |
| 046 | $m_1(s) = \frac{1}{1 + sCR}$ |

For the circuit on the right:

 $v_{out} = v_{in} \frac{R}{R + 1/sC} = v_{in} \frac{sCR}{1 + sRC}$

052 Hence

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$$H_2(s) = \frac{sCR}{1 + sCR}$$

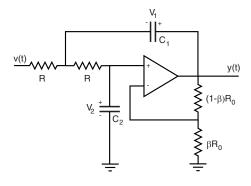


Figure 2: Active low pass filter: Sallen-Key circuit.

1.2 A Second Order System

The circuit shown in Figure 2 is a variation on the Sallen-Key circuit. The circuit contains two capacitors and gives rise to a second order differential equation. For this reason, it is called a second order system. Assuming that R_0 is very large and that $C_1 = C_2$, the circuit transfer function from x to y is:

$$H(s) = \frac{\tau^2/\beta}{s^2 + (3 - 1/\beta)\tau s + \tau^2}$$
(1)

where $\tau = 1/(RC)$ and $0 < \beta \le 1$.

We can parameterize the transfer function of a second order system as follows:

$$H(s) = G \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \tag{2}$$

The two parameters ζ (zeta) and ω_n are called the *damping factor* and the *natural frequency* of the system, respectively, and G is a gain factor.

Questions 1.3

- 1. Find the gain G, natural frequency ω_n , and damping factor ζ of the transfer function H(s) in (2) as functions of β and τ .
- 2. For what value, if any, of β is the circuit critically damped?
- 3. Is the circuit underdamped for any the realizable values of *beta*? If so for what values?
- 4. For what range of $\beta \in (0, 1]$ is this circuit stable and why?
- 5. Derive the transfer function (2) of the Sallen-Key circuit in Figure 2. You can attach this.

¹⁰⁸ 2 Lab Procedure

¹¹⁰ 2.1 Plotting Root Locus¹¹¹

112 A root locus of a transfer function is a plot of the poles of the transfer function (roots of 113 the denominator polynomial) as a scalar parameter is varied. The roots of a polynomial are 114 continuous functions of its coefficients and the coefficients of the denominator polynomial 115 of H(s) are continuous functions of β . Hence as β is varied the roots trace out continuous 116 curves in the complex plane.

117 Create a root locus plot of the poles of the transfer function H(s) of the Sallen-Key circuit 118 as β increases from 0.25 to 1. Do this by creating a loop that cycles through the desired 119 values of β in the range [0.25, 1] and for each value:

- 1. creates a vector of the coefficients of the denominator polynomial den of H(s) (highest to lowest powers of s).
- 2. calls the built-in MATLAB m-file **roots** to compute the roots of the polynomial.
- 3. plots the position of the roots on the complex plane (use hold on and hold off).

To plot a 2-D points with (vectors of) coordinates (x, y) you can use:

127 plot(x,y,'r.','MarkerSize',14);
128

129 Use the grid command:

130 sgrid

131 132 after the end of the loop to add a grid that indicates circles of constant ω_n and radial lines 133 of constant ζ . You can add text to the plot with the command:

134 text(xp,yp,'Root locus for β =1,...,0.25');

Where (xp, yp) is the desired start position for the text. Include such annotation on your plot. Make sure you also label all axis.

138 What does the root locus plot confirm about the stability of this circuit?

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2.2 A Simple Feedback Loop

Now that we have a program to plot a root locus we will use it to explore the stability of the simple negative feedback system shown in Figure 3.

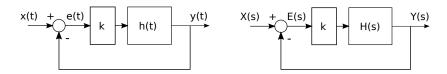


Figure 3: Proportional error feedback regulation. (l) time domain, (r) Laplace domain.

H(s) is the transfer function of the open loop system. It is assumed to be a rational function. The closed loop transfer function from x to y is:

$$F(s) = \frac{kH(s)}{1+kH(s)} = \frac{p(s)}{q(s)}$$

where p(s) and q(s) are polynomials in s. The stability of the system requires that the roots of q(s) be in the open left half of the complex plane (see Laplace transform notes).

For each of the transfer functions H(s) in the list below:

• Compute the closed loop transfer function F(s).

| 162 | • Plot a root locus of the poles of the closed loop system as k varies. |
|------------|--|
| 163 | • Interpret the root locus (explain what is happening). |
| 164 | |
| 165 | • Use the root locus plot to find the largest k (if it exists) such that the poles are stable and sither root of have a demning factor of approximately 0.75 |
| 166 | stable and either real of have a damping factor of approximately 0.75. |
| 167 | 1. $H(s) = \frac{1}{s-1}$ (this system is unstable). |
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| 169 170 | 2. $H(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$ with $\zeta = 0.85$ and $\omega_n = 2$. |
| 171 | 3. $H(s) = \frac{\omega_n^2}{s(s^2 + 2\zeta\omega_n s + \omega_n^2)}$ with $\zeta = 0.85$ and $\omega_n = 2$. |
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| 173 | 4. $H(s)$ as in 3. except with $\zeta = 0.95$ (almost critically damped). |
| 174 | 5. $H(s)$ as in 3. except with $\zeta = 0.65$ (lightly damped). |
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| 176 | Attach all your plots and code to the lab handout prior to handing it in. |
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